

**BJT****Collector current**

$$I_C = \frac{A_E q D_n n_i^2}{N_E W_B} \left( \exp\left(\frac{V_{BE}}{V_T}\right) - 1 \right)$$

$$I_C = I_S \exp\left(\frac{V_{BE}}{V_T}\right)$$

**Base emitter voltage**

$$V_{BE} = V_T \ln \frac{I_C}{I_S}$$

**Thermal Voltage**

$$V_T = 26mV$$

**Transconductance**

$$g_m = \frac{I_C}{V_T}$$

**Output Impedance (Early Effect)**

$$R_O = \frac{V_A}{I_S \exp\left(\frac{V_{BE}}{V_T}\right)} \approx \frac{V_A}{I_C}$$

**Internal Small Signal Resistance**

$$r_\pi = \frac{\beta}{g_m} = \beta \frac{V_T}{I_C}$$

**Terminal Currents**

$$I_E = I_C + I_B$$

$$I_C = \beta I_B$$

**Input Impedances**

Into Base, Emitter AC Grounded

$$r_\pi$$

Into Collector, Emitter AC Grounded

$$\infty$$

Into Collector, Emitter AC Grounded (Early)

$$r_O$$

Into Emitter, Base AC Grounded

$$\frac{1}{g_m}$$

Degenerated CE Stage

$$r_\pi + (\beta + 1)R_E$$

Self Biased Stage

$$r_\pi \parallel \frac{1}{g_m}$$

**Degenerated CE Stage (early)**

$$R_{out} \approx r_O [1 + g_m (R_E \parallel r_\pi)]$$

**Common Base**

Core

$$A_V = g_m R_C$$

With Source Resistance

$$A_V = \frac{R_C}{\frac{1}{g_m} + R_S}$$

Output Impedance (Early)

$$R_{out1} = [1 + g_m (R_E \parallel r_\pi)] r_O + (R_E \parallel r_\pi)$$

$$R_{out} = R_C \parallel R_{out1}$$

Output Impedance of CB/CE are same if under same condition

**Emitter Follower (Common Collector)**

Output Resistance

$$R_{out} = \left( \frac{R_S}{\beta + 1} + \frac{1}{g_m} \right) \parallel R_E \parallel r_O$$

**CMOS****Drain Current**

Saturation No Channel Length Modulation

$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{TH})^2$$

Saturation With Channel Length Modulation

$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{TH})^2 (1 + \lambda V_{DS})$$

Triode Region

$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} [2(V_{GS} - V_{TH})V_{DS} + V_{DS}^2]$$

**Transconductance**

$$g_m = \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{TH})$$

$$g_m = \sqrt{2 \mu_n C_{ox} \frac{W}{L} I_D}$$

$$g_m = \frac{2I_D}{V_{GS} - V_{TH}}$$

**Gain**

$$A_V = g_m R_D$$

**Linear Resistance (Deep Triode Region)**

$$R_{on} = \frac{1}{\mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{TH})}$$

**Channel Length Modulation**

$$R_O = \frac{1}{\lambda I_{DS}}$$

**Regions of Operation**

N-Channel Cut Off

$$V_{GS} \leq V_T$$

N-Channel Linear

$$V_{GS} > V_T, V_{DS} \leq V_{GS} - V_T$$

N-Channel Saturation

$$V_{GS} > V_T, V_{DS} > V_{GS} - V_T$$

P-Channel Cut Off

$$V_{SG} \leq |V_T|$$

P-Channel Linear (Triode)

$$V_{SG} > |V_T|, V_{SD} \leq V_{SG} - |V_T|$$

P-Channel Saturation

$$V_{SG} > |V_T|, V_{SD} > V_{SG} - |V_T|$$

**Common Source**

$$A_v = \frac{-R_D}{\frac{1}{g_m} + R_S}$$

### Current Mirrors

**Output Impedance Bipolar**

$$R_{out} = [1 + g_m(R_E || r_{\pi})]R_o + R_E || r_{\pi}$$

**Output Impedance Bipolar Cascode**

$$R_{out} = [1 + g_{m1}(r_{O2} || r_{\pi1})]r_{O1} + r_{O2} || r_{\pi1} \\ \approx g_{m1}r_{O1}(r_{O2} || r_{\pi1})$$

Maximum limited by beta and ro

$$R_{out,max} = \beta_1 r_{O1}$$

Improved Cascode using PNP current source

$$R_{out} \approx g_{m3}r_{O3}(r_{O4} || r_{\pi3}) || g_{m2}r_{O2}(r_{O1} || r_{\pi2})$$

**Output Impedance CMOS**

$$R_{out} = R_S + (1 + g_m R_S)r_o$$

**Output Impedance CMOS Cascode**

$$R_{out} = (g_{m1}r_{O2})r_{O1} + r_{O2} \\ \approx g_{m1}r_{O1}r_{O2}$$

Improved Cascode using PMOS current source

$$R_{out} \approx g_{m3}r_{O3}r_{O4} || g_{m2}r_{O2}r_{O1}$$

**Short Circuit Transconductance**

$$G_m = \frac{i_{out}}{V_{in}}$$

$$A_v = -G_m R_{out}$$

Cascode Amplifier CMOS

$$G_m = g_{m1}$$

**Current Mirror**

$$I_{copy} = \frac{nI_{REF}}{1 + \frac{1}{\beta}(n+1)}$$

With an extra transistor to reduce error

$$I_{copy} = \frac{nI_{REF}}{1 + \frac{1}{\beta^2}(n+1)}$$

### Differential Amplifiers

**Bipolar Differential Amplifier**

LSA Current

$$I_{C1} = \frac{I_{EE} \exp \frac{V_{in1} - V_{in2}}{V_T}}{1 + \exp \frac{V_{in} - V_{in2}}{V_T}}$$

$$I_{C2} = \frac{I_{EE}}{1 + \exp \frac{V_{in} - V_{in}}{V_T}}$$

Linear Operation Mode

$$|V_{in} - V_{in} | < 4V_T \approx 104mV$$

Small Signal Mode

$$|V_{in} - V_{in} | < 10mV$$

**MOS Differential Pair**

Maximum Differential Input Voltage

$$|V_{in1} - V_{in} |_{max} = \sqrt{2}(V_{GS} - V_{TH})_{equil}$$

Small Signal Mode

$$|V_{in} - V_{in2}| \ll \frac{4I_{SS}}{\mu_n C_{ox} \frac{W}{L}}$$

**Common mode to Differential Mode**

$$A_{CM-DM} = \frac{\Delta R_D}{\frac{1}{g_m + 2R_{SS}}} \approx \frac{\Delta R_D}{2R_{SS}}$$

### Frequency Response

**Miller's Theorem**

$$Z_1 = \frac{Z_F}{1 - A_V}$$

$$Z_2 = \frac{Z_F}{1 - \frac{1}{A_V}}$$

$$C_{in} = (1 - A_v)C_F$$

$$C_{out} = (1 - \frac{1}{A_v})C_F$$

**Transit Frequency**

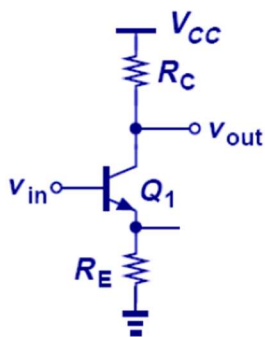
BJT

$$2\pi f_T = \frac{g_m}{C_{GS}}$$

CMOS

$$2\pi f_T = \frac{g_m}{C_\pi}$$

**RECAP: Degenerated CE Stage with Resistive divider**

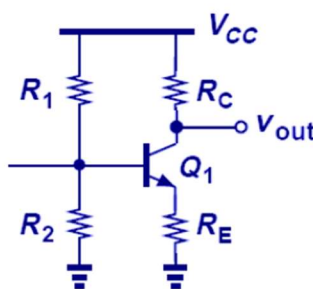


Without series resistance

$$A_v = -\frac{R_C}{\frac{1}{g_m} + R_E}$$

$$R_{in} = r_{\pi} + (\beta + 1)R_E$$

$$R_{out} = R_C$$

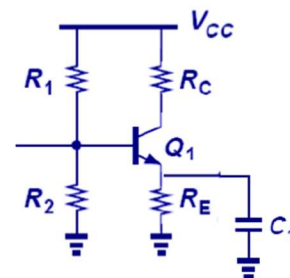


With resistive divider

$$A_v = -\frac{R_C}{\frac{1}{g_m} + R_E}$$

$$R_{in} = [r_{\pi} + (\beta + 1)R_E] \parallel R_1 \parallel R_2,$$

$$R_{out} = R_C$$



With resistive divider & C1

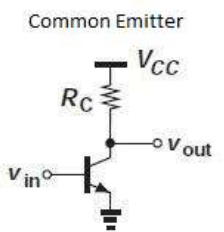
$$A_v = -g_m R_C$$

$$R_{in} = r_{\pi} \parallel R_1 \parallel R_2$$

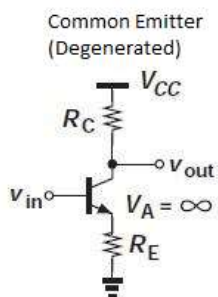
$$R_{out} = R_C.$$

AMPLIFIER TYPE	COMMON BASE	COMMON EMITTER	COMMON EMITTER (Emitter Resistor)	COMMON COLLECTOR (Emitter Follower)
INPUT/OUTPUT PHASE RELATIONSHIP	0°	180°	180°	0°
VOLTAGE GAIN	HIGH $\frac{\alpha R_C}{R_s + r_e}$	MEDIUM $\frac{\beta(R_C \parallel r_o)}{R_s + r_{\pi}}$	MEDIUM $\frac{\beta R_C}{R_s + (\beta + 1)(r_e + R_E)}$	LOW $\frac{(\beta + 1)(R_L \parallel r_o)}{R_s + (\beta + 1)[r_e + (R_L \parallel r_o)]}$
CURRENT GAIN	LOW $\alpha$	MEDIUM $\beta \frac{r_o}{R_C + r_o}$	MEDIUM $\beta$	HIGH $(\beta + 1) \frac{r_o}{r_o + R_L}$
POWER GAIN	LOW	HIGH	HIGH	MEDIUM
INPUT RESISTANCE	LOW $r_e$	MEDIUM $r_{\pi} = (\beta + 1)r_e$	MEDIUM $(\beta + 1)(r_e + R_E)$	HIGH $(\beta + 1)[r_e + (r_o \parallel R_L)]$
OUTPUT RESISTANCE	HIGH $R_C$	MEDIUM $R_C \parallel r_o$	MEDIUM $R_C$	LOW $r_o \parallel \left[ r_e + \frac{R_s}{(\beta + 1)} \right]$

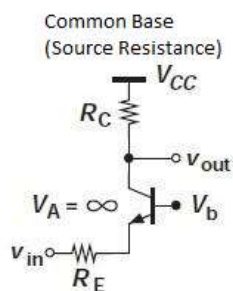
### Voltage Gain Equations



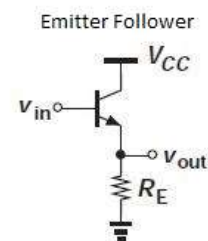
$$A_v = -g_m(R_C \parallel r_o)$$



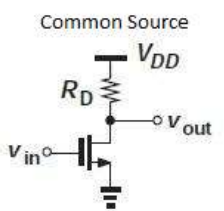
$$A_v = -\frac{R_C}{\frac{1}{g_m} + R_E}$$



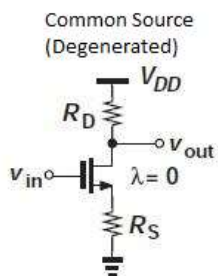
$$A_v = \frac{R_C}{\frac{1}{g_m} + R_E}$$



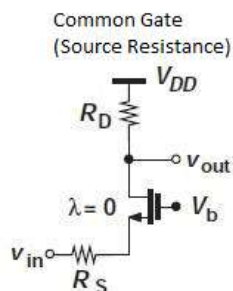
$$A_v = \frac{R_E \parallel r_o}{\frac{1}{g_m} + R_E \parallel r_o}$$



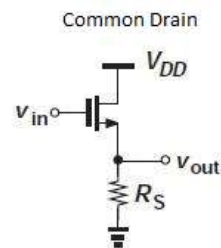
$$A_v = -g_m(R_D \parallel r_o)$$



$$A_v = -\frac{R_D}{\frac{1}{g_m} + R_S}$$



$$A_v = \frac{R_D}{\frac{1}{g_m} + R_S}$$



$$A_v = \frac{R_S \parallel r_o}{\frac{1}{g_m} + R_S \parallel r_o}$$